

A reconstruction algorithm for temporally aliased seismic signals recorded by the InSight Mars lander

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Abstract

The NASA InSight lander successfully placed a seismometer on the surface of Mars. Alongside, a hammering device was deployed that penetrated into the ground to attempt the first measurements of the planetary heat flow of Mars. The hammering of the heat probe generated repeated seismic signals that were registered by the seismometer and can potentially be used to image the shallow subsurface just below the lander. However, the broad frequency content of the seismic signals generated by the hammering extends beyond the Nyquist frequency governed by the seismometer's sampling rate of 100 samples per second. Here, we propose an algorithm to reconstruct the seismic signals beyond the classical sampling limits. We exploit the structure in the data due to thousands of repeated, only gradually varying hammering signals as the heat probe slowly penetrates into the ground. In addition, we make use of the fact that repeated hammering signals are sub-sampled differently due to the unsynchronised timing between the hammer strikes and the seismometer recordings. This allows us to reconstruct signals beyond the classical Nyquist frequency limit by enforcing a sparsity constraint on the signal in a modified Radon transform domain. Using both synthetic data and actual data recorded on Mars, we show how the proposed algorithm can be used to reconstruct the high-frequency hammering signal at very high resolution. In this way, we were able to constrain the seismic velocity of the top first meter of the Martian regolith.

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15 **Key Points:**

- 16 • Hammering of the InSight heat probe generates high-frequency seismic signals that
17 exceed the Nyquist frequency of the seismometer.
- 18 • We developed a new data acquisition and reconstruction workflow that allows for
19 the recovery of the full-bandwidth hammering signals.
- 20 • We thus deliberately turned off the anti-aliasing filters and reconstructed the aliased
21 signal using a sparseness-promoting algorithm.

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Abstract

The NASA InSight lander successfully placed a seismometer on the surface of Mars. Alongside, a hammering device was deployed that penetrated into the ground to attempt the first measurements of the planetary heat flow of Mars. The hammering of the heat probe generated repeated seismic signals that were registered by the seismometer and can potentially be used to image the shallow subsurface just below the lander. However, the broad frequency content of the seismic signals generated by the hammering extends beyond the Nyquist frequency governed by the seismometer’s sampling rate of 100 samples per second.

Here, we propose an algorithm to reconstruct the seismic signals beyond the classical sampling limits. We exploit the structure in the data due to thousands of repeated, only gradually varying hammering signals as the heat probe slowly penetrates into the ground. In addition, we make use of the fact that repeated hammering signals are sub-sampled differently due to the unsynchronised timing between the hammer strikes and the seismometer recordings. This allows us to reconstruct signals beyond the classical Nyquist frequency limit by enforcing a sparsity constraint on the signal in a modified Radon transform domain.

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1 Introduction

The NASA InSight mission successfully landed on Mars in November 2018 (Banerdt et al., 2020). Since then, the SEIS package, consisting of two three-component seismometers (Lognonné et al., 2019), and the heat flow and physical properties package (HP³) (Spohn et al., 2018) were deployed directly onto the surface of Mars. HP³ consists of a self-hammering probe, referred to as the ‘mole’, that penetrates into the shallow subsurface of the Martian regolith with the aim to take thermal conductivity and temperature measurements in order to better understand the Martian planetary heat flow. The hammering mechanism of the mole is designed to slowly dig into the regolith at a rate of about 0.1-1 mm per hammer stroke (Kedar et al., 2017). This means that thousands of repeated hammer strokes are needed to reach the target depth of 5 m.

HP³ hammering generates seismic signals that are recorded by SEIS. These signals can potentially be used to image the shallow subsurface just below the lander (Kedar et al., 2017; Golombek et al., 2018). However, the seismic analysis of the HP³ hammering signals does not address one of the primary mission goals and the experiment was not conceived before finalizing the system design. Therefore, the data acquisition for this opportunistic experiment had to be implemented with the constraints given by the already designed seismic data acquisition flow. Hence, the need to develop the reconstruction workflow discussed in this paper.

SEIS is deployed in close proximity to the HP³ mole at a distance of 1.18 m (Fig. 1). As a result, the travel times of seismic waves generated by the hammering of the mole are extremely short (just a few milliseconds). In order to extract subsurface information from the seismic data (such as seismic velocity and reflectivity), it is thus of crucial importance to have a high temporal resolution for both the recorded seismic signal and the origin time of each mole stroke (i.e., the time the hammer stroke occurs). The latter is known with an accuracy of 1.7 milliseconds from the measurements of an accelerometer that is mounted inside the mole (Spohn et al., 2018). In this paper, we develop a method to additionally increase the temporal resolution of the recorded seismograms beyond the nominal sampling rate of the seismometer. Increasing the temporal resolution is a crit-

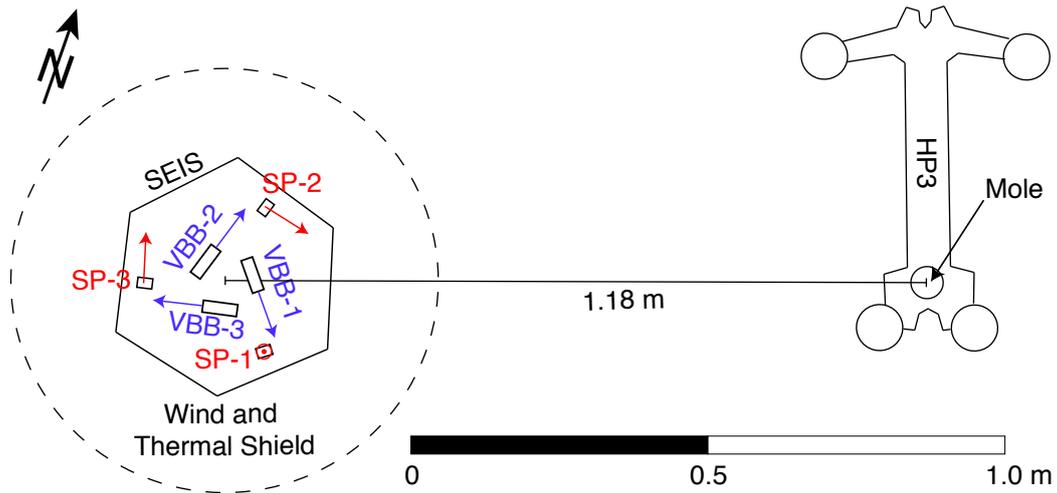


Figure 1. Configuration of SEIS and HP³ on Mars. The orientation and location of the three components of the short-period (SP) and very broadband (VBB) seismometers are marked in red and blue, respectively.

72 ical step since the nominal sampling interval of SEIS is longer than the expected seis-
 73 mic travel time between HP³ and SEIS, effectively preventing the extraction of seismic
 74 velocities (Kedar et al., 2017).

75 SEIS is operated with on-board digital anti-aliasing filters to prepare the seismic
 76 information to be returned to Earth with a maximum sampling rate of 100 samples per
 77 second (sps). This sampling rate provides sufficient temporal resolution for most of the
 78 anticipated Martian seismic signals such as marsquakes and meteorite impacts (Lognonné
 79 et al., 2019; Giardini et al., 2020). However, the impulsive seismic signals generated by
 80 HP³ hammering are very broad-band and may contain frequencies up to and beyond 250 Hz
 81 (Kedar et al., 2017). The application of the nominal anti-aliasing filter would thus re-
 82 sult in a severe loss of information during acquisition. Fig. 2 shows the signal of a single
 83 hammer stroke measured using a commercial seismometer in an analogue experiment
 84 conducted on Earth in the Nevada desert. The pass region of the nominal SEIS anti-aliasing
 85 filter is marked in red. Note how a significant portion of the information including the
 86 dominant signal energy between 100 and 150 Hz would be lost using the nominal anti-
 87 aliasing filter.

88 Given the fact that the idea of using HP³ as a seismic source was conceived after
 89 the implementation of the seismic acquisition hardware, the InSight science team had
 90 to find ways to circumvent limitations of the existing acquisition hardware, such as the
 91 insufficient sampling rate. With the goal to enable the analysis of seismic information
 92 beyond the highest nominal Nyquist frequency of SEIS (i.e., 50 Hz), we designed a data
 93 acquisition and reconstruction workflow that consists of (1) recording aliased data by re-
 94 placing the nominal anti-aliasing FIR filters by all-pass filters and (2) reconstructing the
 95 data at a high sampling rate using a sparseness-promoting algorithm. We illustrate the
 96 success of our method in recovering the high frequency information from the hammer-
 97 ing signals using both synthetic data and actual data from Mars.

98 The HP³–SEIS experiment marks, to the best of our knowledge, the first active seis-
 99 mic experiment ever conducted on Mars (Brinkman et al., 2019). A similar robotic ac-
 100 tive seismic experiment on an extraterrestrial object has only been attempted once be-
 101 fore on the comet 67P Churyumov–Gerasimenko during the Rosetta mission and allowed

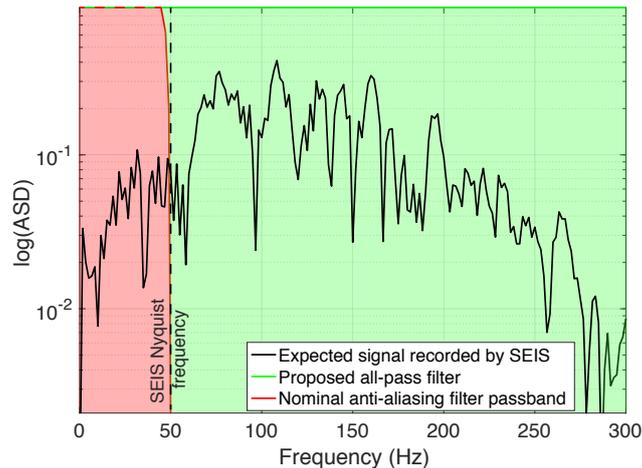


Figure 2. Amplitude spectral density of an HP³ hammering seismic signal obtained in an analogue experiment on Earth. The response of the nominal SEIS anti-aliasing filter is shown in red. The proposed digital all-pass filter passes information throughout the complete bandwidth (green). As a result, the recorded seismic signals will be aliased by several factors when downsampled to 100 sps.

102 for the extraction of the comet’s elastic properties (Knapmeyer et al., 2018). On the lu-
 103 n-
 104 ar surface, the Apollo astronauts conducted active seismic profiling experiments using
 105 mortar and explosive sources (Brzostowski & Brzostowski, 2009), in order to characterise
 106 the shallow subsurface structure at the Apollo 14, 16, and 17 landing sites (Cooper et
 107 al., 1974). In recent years, the Apollo seismic data have been re-processed with modern
 108 analysis tools that allowed for the extraction of novel information on the near-surface
 structure of the Moon (Heffels et al., 2017; Sollberger et al., 2016).

109 2 SEIS data acquisition flow

110 The two seismometers in the SEIS package (VBB and SP) nominally cover a com-
 111 bined seismic bandwidth from 0.01 Hz to 50 Hz (Lognonné et al., 2019). Even though
 112 the instruments would be capable of measuring data at higher frequencies than 50 Hz,
 113 this upper limit is imposed by the maximum sampling rate of the acquisition hardware
 114 (100 sps). The two seismometers record continuously and the data are stored inside a
 115 buffer on-board the lander. From there, the data are first uplinked to the relay satellites
 116 orbiting Mars (usually about two uplink passes per day) and subsequently downlinked
 117 to Earth. Due to the limited storage space of the buffer (64 Gigabit of flash storage) and
 118 data transfer bandwidth limitations, the data volume that can be transferred to Earth
 119 is restricted. The continuous seismic data is therefore down-sampled directly on-board
 120 the lander to a lower sampling rate before it is sent to Earth. Based on the continuous
 121 low-rate data, event data at a higher sampling rate (up to 100 sps) can be requested for
 122 periods of time where seismic signals are observed. In this section, we describe how the
 123 data decimation process is implemented inside the space craft electronics and illustrate
 124 the changes that were implemented for the HP³ hammering experiment to recover the
 125 high-frequency information of the hammering signals.

126 The SEIS signals pass through the data acquisition and decimation flow illustrated
 127 in Fig. 3. The analog voltage signal from the seismometers first passes through an ana-
 128 log anti-aliasing filter, before it is digitised by the sigma-delta analog to digital converter

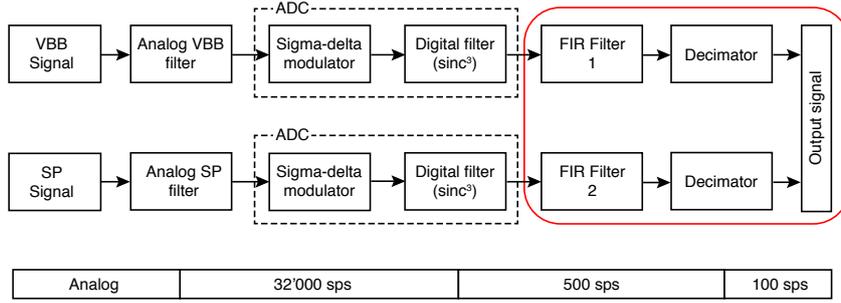


Figure 3. Acquisition and digitization of seismic signals recorded by the two SEIS seismometers. The filtering step in the red box can be changed from Earth by uploading different filter coefficients to the lander. Different filters can be uploaded for VBB and SP (FIR1 and FIR2, respectively)

129 (ADC) on-board the lander at a sampling rate of 32 kHz. Subsequently, the signal is passed
 130 through an additional digital $(\sin(x)/x)^3$ (also called sinc^3) low-pass filter with a cut-
 131 off frequency of 500 Hz and decimated to a sampling rate of 500 sps. The 500 sps sig-
 132 nal is then passed through a digital finite-impulse response (FIR) filter (FIR1 and FIR2
 133 in Fig. 3). Nominally, this filter is set to be a low-pass with a cut-off of 39.8 Hz (-3 dB
 134 half-power point) in order to avoid aliasing in the final 100 sps data product. The FIR
 135 filters of each of the two seismometers can be individually changed by uploading new fil-
 136 ter coefficients to the lander. During HP³ hammering, we replaced the nominal FIR anti-
 137 alias filter on the SP sensor by an all-pass filter (FIR1 in Fig. 3) in order to avoid los-
 138 ing the information above 50 Hz. As a consequence, the decimated signal at 100 sps con-
 139 tains signal up to 500 Hz, which are aliased and thus overlapping the 0-50 Hz range cor-
 140 respondingly.

141 The impulse time and frequency responses of both the nominal (39.8 Hz cut-off)
 142 and the proposed all-pass filters are shown in Fig. 4. Note that the proposed all-pass fil-
 143 ter has a flat frequency response over the full bandwidth. Consequently, its impulse re-
 144 sponse in time corresponds to a single spike. Because the FIR filter coefficients are im-
 145 plemented in the SEIS electronics as signed 32-bit integer numbers, the maximum pos-
 146 sible amplitude of the spike is $(2^{31} - 1)/(2^{32}) \approx 0.5$. As a consequence, the raw data
 147 need to be multiplied with a factor of 2 during the conversion from digital counts to volt,
 148 which results in the loss of 1 bit of resolution (the nominal resolution is 24 bits). Fur-
 149 thermore, the all-pass filter was implemented with a group delay of 0.244 s, whereas the
 150 nominal FIR filter has a group delay of 0.24 s (see delay between the black and the red
 151 curves in the top of Fig. 4).

152 3 Theory

153 The rules dictating the sampling of signals are governed by the Nyquist–Shannon
 154 sampling theorem (Shannon, 1948), stating that in order to reconstruct a signal from its
 155 samples, the signal must contain no information at and above the Nyquist frequency cor-
 156 responding to half the sampling frequency. However, the Nyquist–Shannon sampling the-
 157 orem assumes sampling of a single quantity of the underlying signal. If multiple data types,
 158 corresponding to data filtered before sampling with linearly independent filters in the
 159 domain of sampling are available, then the Nyquist–Shannon sampling criterion is re-
 160 laxed proportionally to the new degrees of freedom added to solve the problem. This so-
 161 called generalized sampling theorem (Papoulis, 1977) provides the mathematical frame-
 162 work for the reconstruction algorithm we propose in this paper.

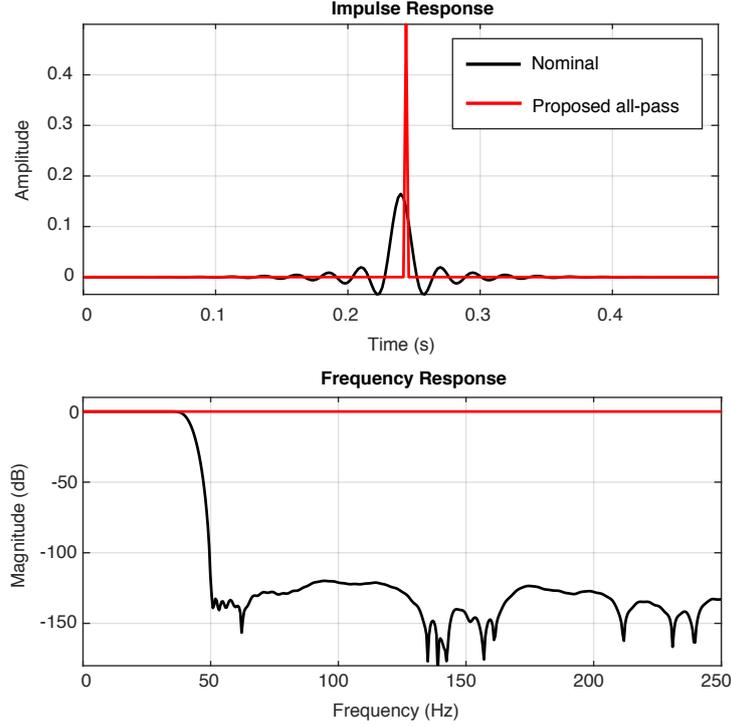


Figure 4. Impulse responses and frequency responses of the digital FIR filters implemented in the SEIS acquisition electronics.

163 In case of the HP³ hammering signals, we have access to multiple realizations of
 164 approximately the same signal from subsequent mole strokes. Because the source trig-
 165 gering and sampling process are unsynchronized, the different realizations will appear
 166 as if they have been filtered in time with different Fourier shift filters (i.e., the 100 sps
 167 sampling comb is randomly shifted in time for each hammering signal). While the use
 168 of the generalized sampling theorem as described above relies on multiple realizations
 169 of the same signal, we allow for the reconstruction of smoothly varying signals by exploit-
 170 ing the inherent linear data structure when the hammer recordings are rearranged into
 171 a 2D signal (with time relative to the hammer stroke on one axis and space on the other),
 172 causing the signal to have a sparse representation in the Radon transform domain.

173 Reconstruction problems are inherently underdetermined (i.e., the number of sam-
 174 ples that are sought to be recovered is always greater than the number of data points
 175 that are available to constrain the problem). Such problems thus need to be regularized
 176 in some way, which means that a priori information about the signal must be included
 177 to achieve a successful reconstruction. Recent advances in signal processing make use
 178 of signal sparsity as a priori knowledge to regularize the underdetermined reconstruc-
 179 tion problem (Candès et al., 2006a,b; Donoho, 2006). Sparsity is thereby usually described
 180 either by the l_0 - or the l_1 -norm of the signal and penalties are given to reconstructions
 181 with high l_0 - or l_1 -norm. A prerequisite is that the signal has a sparse representation
 182 in some transform domain and the success highly depends on the compressibility of the
 183 signal and thus the selection of the sparsifying transform (i.e., an operator mapping the
 184 signal data vector to a sparse vector). The concept of sparsity-constrained reconstruc-
 185 tion has been successfully applied, for example, to accelerate magnetic resonance imag-
 186 ing (Lustig et al., 2007) or to interpolate seismic data (Herrmann & Hennenfent, 2008).

187 Here, we devise a signal reconstruction algorithm using sparsity constraints. The
 188 key characteristics of the HP³ seismic signals that are exploited for reconstruction are:

- 189 1. The hammering signal is highly repeatable and only slowly varying in space (depth)
 190 due to the slow penetration rate of the mole.
- 191 2. The signal sample times of repeated hammering signals are different since the trig-
 192 ger time of the hammer mechanism is unsynchronised with the sampling process
 193 of SEIS.

194 As we will demonstrate in the following, these characteristics have the effect that the ham-
 195 mering signals are highly compressible using a modified Radon transform. This prop-
 196 erty, in addition with the quasi-random sub-sampling of the signal due to the unsynchro-
 197 nised timing between the hammer strokes and the recording system provides the founda-
 198 tion for successful sparse reconstruction.

199 3.1 Signal compressibility

Let $d(t, x)$ be a 2D signal (e.g., seismic data) of time variable t and space variable x . The linear Radon transform allows for representing the signal as a superposition of integrals over straight lines (Radon, 1917). Each point in the transform domain (in the following referred to as τ - p -plane) then corresponds to the line integral of $d(t, x)$ over the straight line with intercept time τ and slope (or slowness) p . Here, we begin with the inverse Radon transform (i.e., the operation corresponding to the summation of all points passing through a line). It can be formulated in the following way

$$d(t, x) = \mathcal{R}^* m_\delta(\tau, p) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} m_\delta(\tau, p) \delta(t - \tau - px) d\tau dp, \quad (1)$$

200 where $m_\delta(\tau, p)$ is the representation of the signal in the τ - p -plane, \mathcal{R}^* is the inverse Radon
 201 transform operator, and $\delta(t - \tau - px)$ is the basis function of the transform describing
 202 lines of slope p and intercept time τ .

203 If the signal $d(t, x)$ shows an underlying 2D linear structure, it will focus at sparse
 204 locations in the Radon transform domain, since the transform compresses each line to
 205 a point (i.e., the Radon transform is a sparsifying transform for such a signal).

206 In the following, we assume that $d(t, x)$ is a seismic signal. Seismic data are always
 207 band-limited due to the blurring effect caused by the band-limited source wavelet. This
 208 reduces the temporal focusing capabilities and thus the sparsifying potential of the con-
 209 ventional Radon transform for seismic data. A sparser τ - p -representation of the data can
 210 be found when information on the seismic wavelet (i.e., the source-time function of the
 211 seismic source) is included into the basis function of the transform (Gholami, 2017).

212 Let $w(t)$ be a suitably defined, known wavelet that is a reasonable approximation
 213 to the actual source time function of the seismic source. We now modify the basis func-
 214 tion of the Radon transform to find a sparser τ - p -plane representation of the signal by
 215 including information on the wavelet (Gholami, 2017). The modified basis function now
 216 reads:

$$w(t - \tau - px) = w(t) * \delta(t - \tau - px). \quad (2)$$

217 This new basis function is still constant along all lines of slope p but at a fixed point
 218 in space x , it is a wavelet shifted in time. This allows for a particularly good represen-
 219 tation of seismic signals as a super-position of band-limited transient plane waves. It can
 220 be shown that this modified Radon transform can simply be expressed by the conven-

221 tional Radon transform and an additional deconvolution with the wavelet $w(t)$. For the
 222 inverse of this modified Radon transform, it follows that (Gholami, 2017):

$$\begin{aligned}
 d(t, x) &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} m_w(\tau, p) w(t - \tau - px) d\tau dp \\
 &= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} m_w(\tau, p) [w(t) * \delta(t - \tau - px)] d\tau dp \quad (3) \\
 &= w(t) * \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} m_w(\tau, p) \delta(t - \tau - px) d\tau dp = w(t) * \mathcal{R}^* m_w(\tau, p),
 \end{aligned}$$

223 where $m_w(\tau, p)$ are the τ - p -coefficients of the signal in the modified Radon transform do-
 224 main. Eq. 3 makes the implementation straightforward as it allows one to use existing
 225 Radon transform routines. In the following, we make use of a recently published, fast
 226 implementation of the Radon transform (Andersson & Robertsson, 2019).

227 3.1.1 Discrete implementation

228 For the discrete implementation of this modified Radon transform, let $\mathbf{d} \in \mathbb{R}^M$,
 229 and $\mathbf{m} \in \mathbb{R}^N$ be vectors containing discrete samples of the signal coefficients in the t -
 230 x - and τ - p -planes, respectively. The number of discrete samples are given by $M = n_t n_x$,
 231 and $N = n_\tau n_p$, with n_t being the number of time samples, n_x the number of samples
 232 in space, n_τ the number intercept times, and n_p the number of slowness values. In the
 233 following $\|\cdot\|_p$ is the ℓ_p -norm of a vector and (using the example of \mathbf{m}) is defined as $\|\mathbf{m}\|_p :=$
 234 $\left(\sum_{i=1}^M |m_i|^p\right)^{\frac{1}{p}}$. The discrete forward Radon transform can now be formulated in the
 235 form of an optimization problem based on Eq. (3) to find the best-fitting (in a least-squares
 236 sense) τ - p -representation $\hat{\mathbf{m}}$ of the signal as:

$$\hat{\mathbf{m}} = \arg \min_{\mathbf{m}} \|\mathbf{d} - \mathbf{W}\mathbf{R}^*\mathbf{m}\|_2. \quad (4)$$

237 Here, $\mathbf{W} \in \mathbb{R}^{M \times M}$ is a block-diagonal matrix with n_x blocks, each block corre-
 238 sponding to a Toeplitz matrix $\mathbf{T} \in \mathbb{R}^{n_t \times n_t}$ that is constructed from the wavelet by cyclic
 239 permutation. Left multiplication with \mathbf{W} corresponds to a convolution with the wavelet.
 240 The matrix $\mathbf{R} \in \mathbb{R}^{M \times N}$ is the Radon transform matrix, which is readily implemented
 241 in the frequency domain with the elements given by $R_{jk} = e^{i\omega p_j x_k}$, where ω is the an-
 242 gular frequency. The asterisk marks the Hermitian conjugate operator. The solution of
 243 the optimization problem typically requires some form of stabilization, such as Tikhonov
 244 regularization.

245 The improvements in signal compressibility that can be achieved using the mod-
 246 ified, sparse Radon transform compared to the conventional Radon transform are illus-
 247 trated in Fig. 5. A synthetic signal is shown that comprises two band-limited plane waves,
 248 the first with slowness $p_1 = 0$ s/m and intercept time $\tau_1 = 0.15$ s and the second with
 249 slowness $p_2 = 0.25$ s/m and intercept time $\tau_2 = 0.16$ s (Fig. 5a). The conventional, lin-
 250 ear Radon transform focusses the two waves at the expected locations in the τ - p -plane
 251 (Fig. 5b). Note that the temporal resolution is limited due to the sub-optimal choice of
 252 the basis function. Additionally, the energy of the two events smears out due to the lim-
 253 ited aperture of the data in the space direction. The modified Radon transform accounts
 254 for the band-limited nature of the data and allows to effectively compress each plane wave
 255 to a single point in τ - p -space (Fig. 5c).

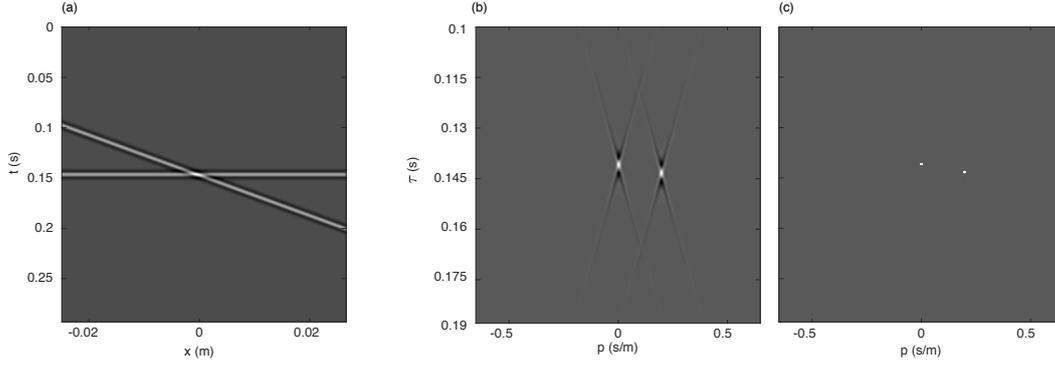


Figure 5. (a) Two linear band-limited events in the space-time domain. (b) Conventional linear Radon transform of the data in (a). (c) Sparse, modified Radon transform.

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3.2 Signal reconstruction

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The reconstruction problem can be understood as a modified version of the forward Radon transform with an additional sparsity constraint. Instead of having access to the fully sampled data \mathbf{d} , we only have access to the sub-sampled data $\mathbf{b} \in \mathbb{R}^P$, where P is the number of sub-samples ($P \ll M$). For the specific problem of reconstructing the HP³ seismic signals, \mathbf{b} is under-sampled in time and thus shows pronounced aliasing. In order to reconstruct \mathbf{d} , we need to solve an underdetermined optimization problem. Here, we formulate the signal reconstruction problem in the form of the following basis pursuit denoise problem (BPDN), seeking for the sparsest set of τ - p -coefficients that explains the data with a misfit smaller than σ (an estimate of the noise level in the data) by ℓ_1 -norm minimization:

$$\hat{\mathbf{m}} = \arg \min_{\mathbf{m}} \|\mathbf{m}\|_1 \quad s.t. \quad \|\mathbf{b} - \mathbf{GWR}^* \mathbf{m}\|_2 \leq \sigma. \quad (5)$$

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Here, the matrix $\mathbf{G} \in \mathbb{R}^{P \times M}$ is the sampling operator selecting those samples from the model that are contained in the observed data \mathbf{b} . \mathbf{G} can be easily constructed from the identity matrix by deleting rows corresponding to samples that are not included in \mathbf{b} . We use the solver SPGL1 (van den Berg & Friedlander, 2009, 2011), which allows for an efficient solution of the BDPN problem by breaking it down into a series of so-called LASSO problems, each of the form

$$\hat{\mathbf{m}} = \arg \min_{\mathbf{m}} \|\mathbf{b} - \mathbf{GWR}^* \mathbf{m}\|_2 \quad s.t. \quad \|\mathbf{m}\|_1 \leq \rho_k, \quad (6)$$

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where ρ_k is the ℓ_1 -norm constraint for the solution of the k^{th} LASSO problem (k being the iteration counter). For a well-defined series of constraints $\rho_0 < \rho_1 < \dots < \rho_k$, the solution converges to the solution of the BDPN problem (Eq. 5), as soon as the least-squares misfit reaches the pre-defined error level σ . It turns out that the series of ℓ_1 -norm constraints ρ_k can be readily defined using a Newton root-finding method on the Pareto curve (van den Berg & Friedlander, 2009). The Pareto curve traces the optimal trade-off between the least-squares misfit and the ℓ_1 -norm of the solution. It is convex, decreasing and continuously differentiable. Each solution of the k LASSO problems lies on the Pareto curve and the slope of the curve at that point can be expressed in closed form (van den Berg & Friedlander, 2009). This property is used to find the optimal ℓ_1 -constraint for the next LASSO problem using Newton's method. At each iteration, the new LASSO problem can be 'warm-started' using the solution of the previous iteration. For details

285 on this procedure, we refer the reader to (van den Berg & Friedlander, 2009, 2011) and
 286 Appendix B in (Lin & Herrmann, 2013). After convergence, the reconstructed signal $\hat{\mathbf{d}}$
 287 can be found by $\hat{\mathbf{d}} = \mathbf{WR}^* \hat{\mathbf{m}}$.

Algorithm 1: Reconstruction of HP³ hammering seismic signals.

Result: Sparse τ - p representation of the reconstructed signal $\hat{\mathbf{m}}$
Input: Aliased seismic data \mathbf{b} , target data misfit σ , minimum medium velocity
 c_0 .

- 1 Initialize iteration counter $k \leftarrow 0$;
- 2 Initialize l_1 -norm constraint $\rho_0 \leftarrow 0$;
- 3 Initialize \mathbf{m}_0 as zero vector;
- 4 **while** $\|\mathbf{b} - \mathbf{GWR}^* \mathbf{m}_k\|_2^2 \geq \sigma$ **do**
- 5 $\rho_{k+1} \leftarrow$ determine from σ and ρ_k using Newton's method on the Pareto
 curve;
- 6 $\mathbf{m}_{k+1} \leftarrow \|\mathbf{b} - \mathbf{GWR}^* \mathbf{m}\|_2^2$ s. t. $\|\mathbf{m}\|_1 < \rho_{k+1}$;
- 7 $k \leftarrow k + 1$;
- 8 **end while**
- 9 Reconstruct signal as $\hat{\mathbf{d}} = \mathbf{WR}^* \hat{\mathbf{m}}$

288 There are three user-specified input parameters for the reconstruction algorithm:
 289 (1) the target data misfit σ (i.e., the noise level in the data), which can be estimated di-
 290 rectly from the data during periods where the hammer is not active, (2) the source wavelet,
 291 and (3) the slowness range that is used to parameterize the Radon transform. This slow-
 292 ness range is naturally bounded by the lowest seismic velocities in the medium, which
 293 are typically shallow S-wave velocities. In the τ - p plane, all signal must thus be contained
 294 in the cone-shaped, convex set $\mathcal{C} = \left\{ (\tau, p) : |p| \leq \frac{1}{c_0} \right\}$, where c_0 is the lowest seismic
 295 velocity in the medium. This puts an additional constraint on the reconstructed signal
 296 (i.e., it must only have support within \mathcal{C}). Everything outside the set \mathcal{C} corresponds to
 297 noise. On Mars, the shallow near-surface seismic shear wave velocities are expected to
 298 be very low, on the order of $c_0 = 40$ -50 m/s (Morgan et al., 2018). The proposed re-
 299 construction algorithm is summarized in Algorithm 1.

300 4 Numerical example

301 In order to illustrate the reconstruction algorithm, we generated synthetic data us-
 302 ing a time-domain finite-difference method for heterogeneous elastic media (Virieux, 1986).
 303 We used a near-surface velocity model that is based on mechanical tests conducted on
 304 regolith simulants in the laboratory (Delage et al., 2017; Morgan et al., 2018). Addition-
 305 ally, we added 2D stochastic velocity fluctuations based on a Von Kármán model (Korn,
 306 1993; Goff & Holliger, 2003) in order to simulate a heterogeneous subsurface. For illus-
 307 tration, the P-wave velocity distribution of the final model is given in Fig. 6.

308 Making use of source-receiver reciprocity, we then generated synthetic seismic data
 309 for a total of 1000 mole positions in a single computation by placing a vertically directed
 310 force source at the location of SEIS (marked by a red asterisk in Fig. 6a) and 1000 re-
 311 ceivers spaced vertically at 5 mm from the surface down to a depth of 5 m at a lateral
 312 offset from SEIS of 1.5 m at the surface (receivers marked by the black line in Fig. 6a).
 313 We used an experimentally-determined source-time function from an analogue experi-
 314 ment on Earth with a dominant frequency of about 150 Hz. We then interpolated the
 315 computed data to a receiver spacing of 1 mm in order to emulate the actual penetration
 316 rate of the mole. Finally, we concatenated all of the resulting 5000 hammering signals
 317 to a single, continuous record. The time differences between individual hammer strokes

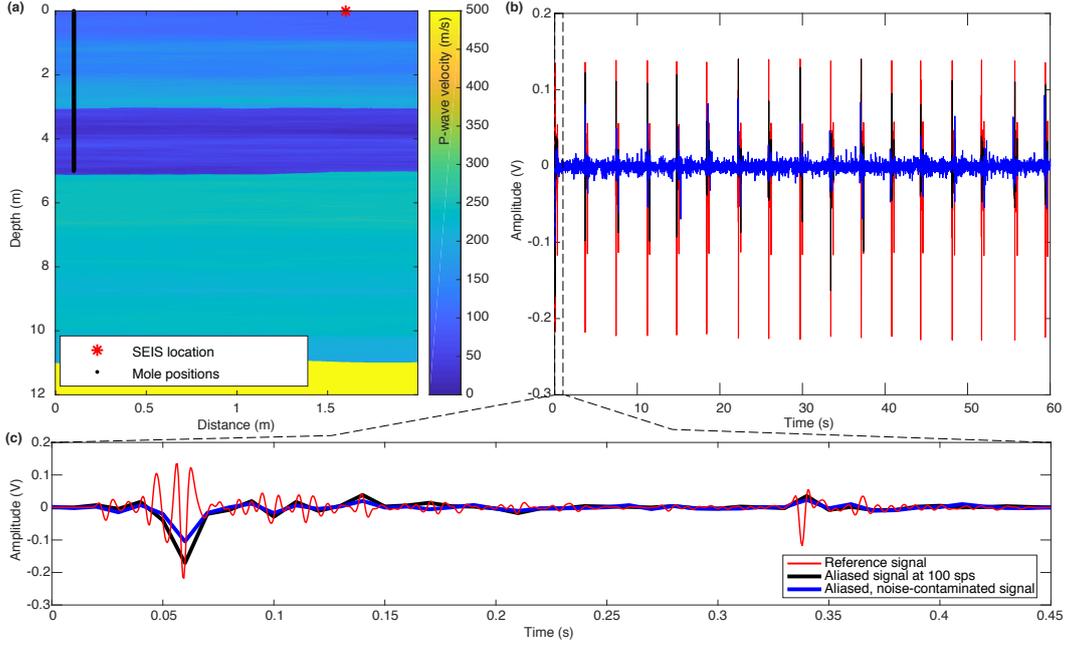


Figure 6. (a) Input velocity model for synthetic data computation. (b) Synthetic dataset emulating 60 seconds of HP³ hammering recorded by SEIS. (c) Zoom-in into the first hammer stroke.

318 were chosen from a normal distribution with a mean value of 3.7 s and a standard de-
 319 viation of 0.1 seconds to mimic the real duration and variations of the mole’s hammer-
 320 ing cycle (Spohn et al., 2018).

321 The first 60 seconds of the resulting record are shown in Fig. 6b. A zoom-in show-
 322 ing the first hammer stroke is provided in Fig. 6c. The red line marks the unaliased data
 323 sampled at 32 kHz before it would pass through the down-sampling flow on-board the
 324 lander (see Fig. 3). The black line marks the same signal after passing through the on-
 325 board acquisition flow using the proposed all-pass FIR filter in the final step (see Fig. 4)
 326 before decimating the signal to 100 sps. Note that the signal is now severely undersam-
 327 pled (aliased). We then additionally added noise to the signal giving the signal marked
 328 by the blue line in Figs 6b-c, which now corresponds the final input that we used to test
 329 the proposed reconstruction algorithm. The added noise corresponds to actual noise that
 330 was measured on Mars during an early phase of the InSight mission with the proposed
 331 all-pass FIR filter on the SP sensor.

332 For reconstruction, we then sorted the data into a 2D matrix, where each column
 333 corresponds to a single hammer stroke signal (Fig. 7). Note that the zero-time corresponds
 334 to the time when the hammer strike occurs. This zero-time time can be retrieved with
 335 an accuracy of $1.7e-3$ s from the measurements of an accelerometer that is mounted in-
 336 side the mole (Spohn et al., 2018). The left panel in Fig. 7 shows the assembled data ma-
 337 trix of the unaliased reference signal at a sampling rate of 2000 sps. For the test, we only
 338 use 10 minutes of data (160 hammer strokes). Note that the signal is only slowly vary-
 339 ing with depth (due to the slow penetration rate of the mole and the repeatability of the
 340 hammering signal), resulting in the linear structure that is exploited by the proposed re-
 341 construction algorithm.

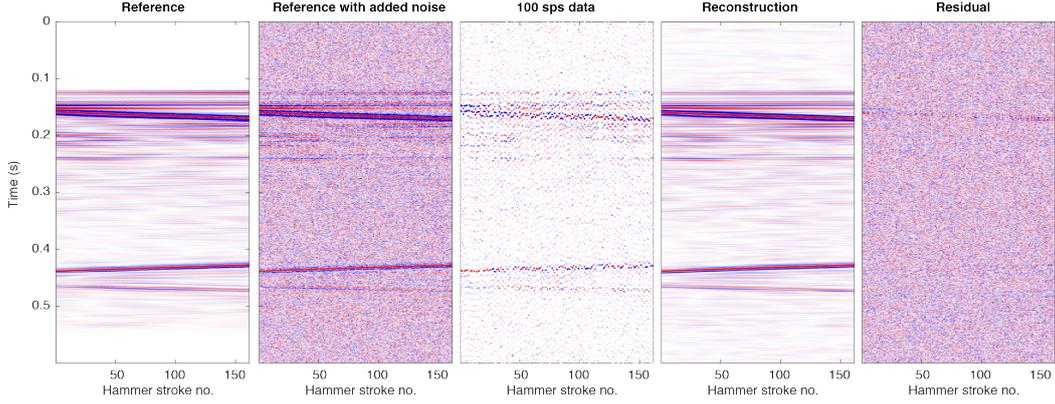


Figure 7. Application of the proposed reconstruction algorithm to a synthetic test dataset (see text for details).

342 The second panel from the left shows the reference data with added real noise as
 343 measured on Mars with the all-pass FIR filter. The samples contained in the 100 sps,
 344 aliased data (input to the reconstruction algorithm) are given in the third panel. Note
 345 that even though the signal is regularly sampled in time (at 100 sps), the sampling in
 346 2D appears to be close to random. This is due to the fact that the timing of the ham-
 347 mer strokes is not synchronised with the SEIS recording system. The respective sub-sampling
 348 of each hammer signal depends on the duration of the hammer cycle (subject the small
 349 variations caused by ambient conditions) and the relative positions of the mole and SEIS.
 350 As a result, each repeated signal is sub-sampled differently, resulting in the random 2D
 351 sampling pattern, which provides an optimal basis for the proposed reconstruction al-
 352 gorithm using the generalized sampling theorem (Papoulis, 1977).

353 We then estimated an average source-time function from the aliased data, by com-
 354 bining the samples of 20 neighbouring hammer stroke signals to a single trace at 2000 sps,
 355 from which we extracted a wavelet by time-windowing the first-arrival.

356 The output of the proposed reconstruction algorithm (reconstructed to a sampling
 357 rate of 2000 sps) is shown in the fourth panel in Fig. 7. For the parametrisation of the
 358 Radon transform, we used slowness values ranging from -0.04 s/m to $+0.04$ s/m (re-
 359 construction is limited to events with a minimum absolute apparent velocity greater than
 360 25 m/s). The high-frequency signal is accurately retrieved by the reconstruction algo-
 361 rithm. Note that random noise appears to be suppressed in the output compared to the
 362 noise-contaminated input data. This is a positive side-effect of the proposed reconstruc-
 363 tion approach owing to the properties of the Radon transform. The integration along
 364 straight lines will cause coherent energy (signal) to add constructively and focus in the
 365 Radon domain, while random noise tends to spread out over the whole domain and can-
 366 cel out. By promoting sparsity of the signal in the Radon domain, the signal is effectively
 367 denoised since only the largest coefficients (corresponding to signal) are kept in the re-
 368 construction. The rightmost panel in Fig. 7 shows the reconstruction residual (i.e., the
 369 difference between the reference and the reconstructed signal). Note that the residual
 370 is mainly dominated by noise, indicating that the underlying signal was successfully re-
 371 constructed. Some minor reconstruction errors seem to be present at the edges for the
 372 events with the lowest apparent velocity. These errors are likely Radon transform arti-
 373 facts (linear flares) caused by the truncation of the dataset (Andersson & Robertsson,
 374 2019).

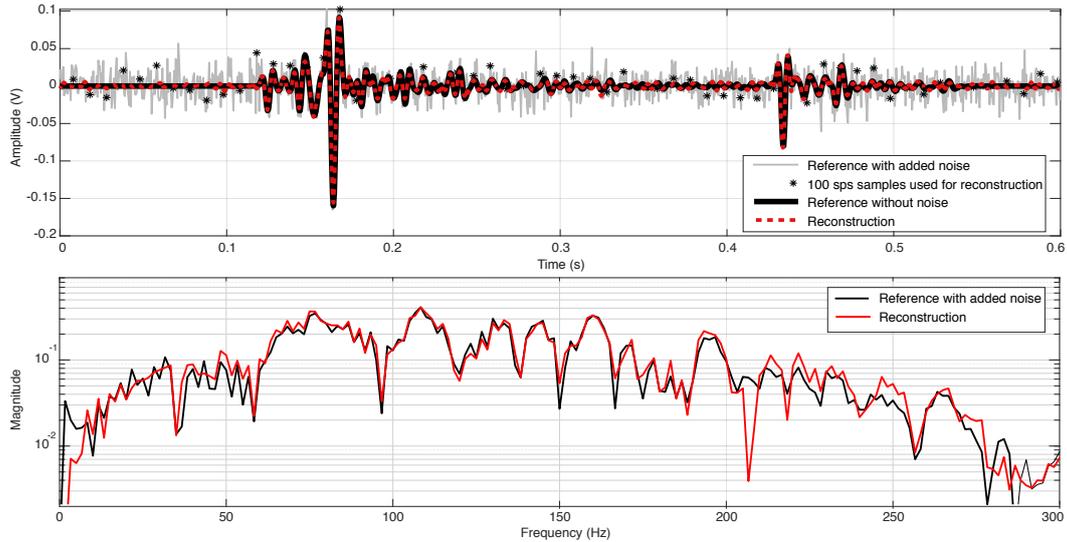


Figure 8. Reconstruction result illustrated on a single hammering signal (hammer stroke no. 60 in Fig. 7). Top: Time domain result. Bottom: Frequency domain result.

375 In order to further illustrate the performance of the reconstruction algorithm, we
 376 provide the result for a single hammering signal in Fig. 8. The top panel shows the re-
 377 sult in the time domain. The black line corresponds to the fully sampled reference sig-
 378 nal. The noise-contaminated samples at 100 sps that are used for the reconstruction are
 379 marked by asterisks. The reconstructed signal is plotted in red. Note that the recon-
 380 struction result is close to the noise-free reference signal (black). An inspection of the ampli-
 381 tude spectrum (bottom panel) confirms that the reconstruction appears to recover the
 382 underlying signal throughout the entire signal bandwidth.

383 4.1 Sensitivity on the source wavelet

384 The task of directly estimating the source wavelet from the data can be challeng-
 385 ing in cases where the signal is dominated by strong resonances that lead to a quasi-monochromatic
 386 appearance of the data. Additionally, in certain cases the waveform of first-arriving wave
 387 does not accurately represent the source-time function (e.g. due to interference of dif-
 388 ferent arrivals). It is thus critical to evaluate how much the reconstruction results suf-
 389 fer from a poorly estimated wavelet. To address this issue, we performed a sensitivity
 390 analysis using the synthetic data set described above (noise-free version).

391 Reconstruction results are shown in Fig. 9 in comparison to the reference for dif-
 392 ferent strategies of choosing the wavelet basis: (a) the wavelet is directly estimated from
 393 the aliased data by combining samples from neighbouring traces as described above, (b)
 394 the wavelet is pre-described by a Ricker wavelet with a center-frequency corresponding
 395 to the actual dominant frequency in the data (150 Hz), (c) the wavelet is pre-described
 396 by a Ricker wavelet with an overestimated center-frequency (200 Hz), and (d) the wavelet
 397 is simply set to a Dirac delta function. The first three approaches (a)-(c) all yield almost
 398 identical results with a residual reconstruction error smaller than 1 percent compared
 399 to the ground truth. Thus, a slight error in the estimation of the wavelet only has a mi-
 400 nor impact on the reconstruction results. Choosing a Dirac delta function as wavelet ba-
 401 sis clearly leads to poorer results (reconstruction error of about 10 percent). Neverthe-
 402 less, a more suitable wavelet can easily be found from such an initial result by Wiener

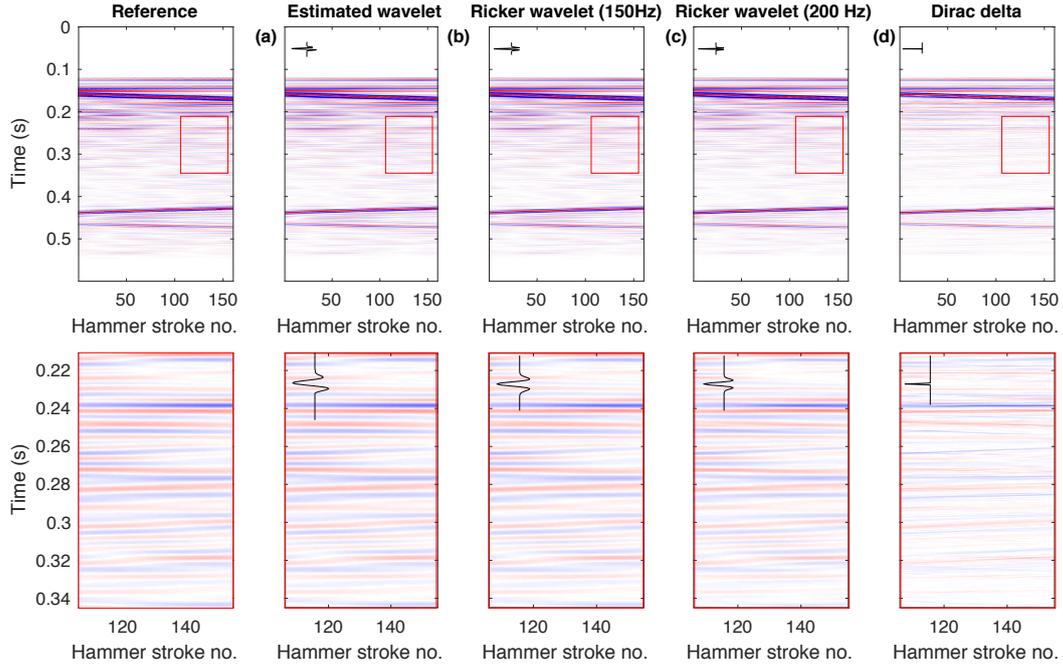


Figure 9. Impact of the user-prescribed source wavelet on the reconstruction result. Top row: Full data set. Bottom row: Zoom-in into the area labeled with a red box. The wavelet used for reconstruction is shown in the top left corner. The shown reconstruction results are obtained using (a) a wavelet estimated directly from the aliased data (see text for details), (b) a Ricker wavelet with the correct center frequency of the signal of 150Hz, (c) a Ricker wavelet with an over-estimated center-frequency of 200 Hz, and (d) a Dirac delta function.

403 deconvolution, as proposed by Gholami (2017). The wavelet can be iteratively adapted
404 until no change in the reconstruction result is observed.

405 The relatively minor impact of the wavelet on the reconstruction quality can be
406 explained by the way the data is compressed by the Radon transform. The major con-
407 tribution to the compression comes from the mapping of near-horizontal (slowly-varying)
408 features in the horizontal (spatial) direction to points in the Radon transform domain.
409 In comparison, the compression of features in the temporal direction due to the choice
410 of the wavelet basis only amounts to a minor contribution of the overall compression rate.

411 5 Mars data example

412 We applied the proposed reconstruction algorithm to actual signals recorded on Mars.
413 The HP³ mole began its hammering operations on Mars on February 28, 2019. After about
414 the first five minutes of hammering (≈ 80 strokes), the mole got stuck at a depth of about
415 30 centimeters and did not make any significant progress in depth anymore. The cause
416 of this is currently still under investigation. In an attempt to recover the mole and to
417 extract diagnostics on the cause of the encountered anomaly, the mole has in the mean-
418 time conducted close to 10'000 hammer strokes. All strokes were recorded by both SEIS
419 seismometers with a high signal-to-noise ratio.

420 We apply our reconstruction algorithm to data from a short hammering session,
421 consisting of 200 hammer strokes (about 12 minutes of hammering) carried out on Mars
422 on March 26, 2019. During this hammering session, the SP sensor was operated using
423 the proposed all-pass FIR filter (Fig. 4) while the VBB sensor was operated with the nom-
424 inal anti-aliasing filter (no signal above 50 Hz recorded). Due to the encountered prob-
425 lems, the mole did not make any noticeable progress in depth during the 200 hammer
426 strokes.

427 The data are characterised by a high signal-to-noise ratio on both the SP and VBB
428 sensor. The VBB data confirmed that the hammering signal is highly repeatable. The
429 aliased, 100 sps signals (recorded on the SP sensor) of all 200 strokes arranged in a 2D
430 matrix are shown in the left panel in Fig. 10. Since the accelerometers mounted inside
431 the mole need to be calibrated and did not provide sufficiently precise information on
432 the trigger time of the hammer strokes for the first few hammer sessions, we had to rely
433 on a different approach to align the data: We first upsampled the 0-50 Hz data from the
434 VBB sensor to 2000 sps and then used a cross-correlation procedure to align the indi-
435 vidual hammer stroke signals. This procedure allowed us to find the relative shifts of the
436 100 sps subsampling comb function from stroke to stroke, which we used to determine
437 the subsampling operator (matrix \mathbf{G} in Eq. 5). Note that, as a result of this procedure,
438 the zero-time in Fig. 10 does not correspond to the actual hammering time. For later
439 sessions, we could directly use the calibrated trigger time from the mole.

440 The reconstruction result for the 200 hammer strokes is given in the second panel
441 in Fig. 10. As expected, the signal characteristics do not significantly change between
442 different hammer strokes. Differences in the signal at later times (later than 0.15 s) are
443 likely caused by variations in the timing of the second and third sub-stroke of the ham-
444 mering mechanism (Spohn et al., 2018). The frequency spectra displayed in the right-
445 most panel of Fig. 10 illustrate that the signal contains a significant amount of informa-
446 tion above the original Nyquist frequency of 50 Hz. Note that this information would
447 have been lost using the nominal anti-aliasing filters. The low-frequency portion of the
448 signal is dominated by long-lasting reverberations following the first arrival. These re-
449 verberations have a dominant frequency of about 25 Hz, as can be seen from the frequency
450 spectra in the right panel of Fig. 10 (distinct peak at 25 Hz for each stroke). The suc-
451 cessful reconstruction of these reverberation illustrate that also quasi-monofrequent sig-

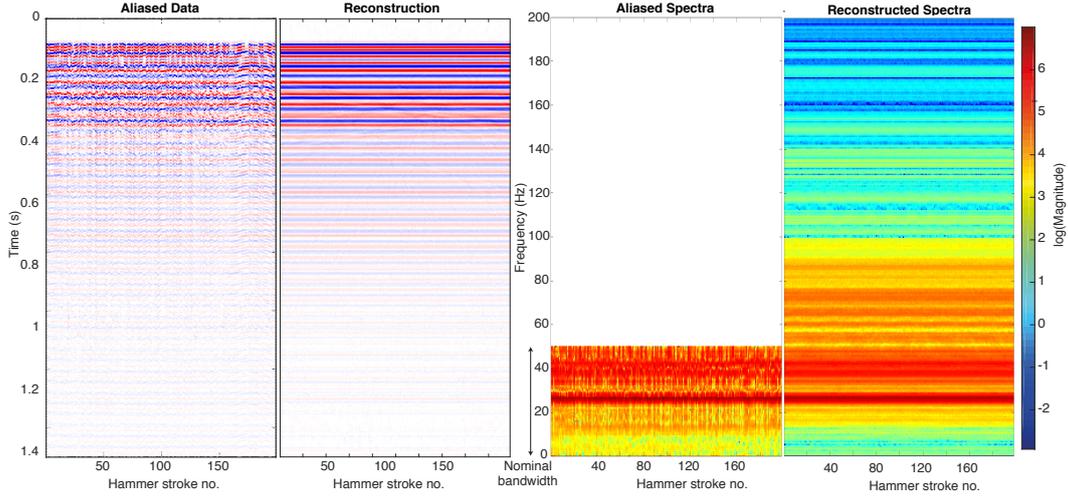


Figure 10. Application of the proposed signal reconstruction algorithm to actual data recorded on Mars. Left two panels: Time domain result. Right: Frequency domain result.

452 nals can be recovered well by our algorithm. The cause of the reverberations is currently
 453 under investigation.

454 5.1 Preliminary results on Martian near-surface properties

455 The high-sampling rate data that we obtained by applying the proposed reconstruction
 456 algorithm allowed us to successfully estimate the P-wave velocity in the top first me-
 457 ter of the Martian regolith. The travel time of the first-arriving wave was determined
 458 to be 9.40 ± 2.68 milliseconds over a distance of 1.11 m (with the mole tip at a depth of
 459 33 cm pointing towards SEIS) resulting in a P-wave velocity of 118 ± 34 ms^{-1} (Lognonné
 460 et al., 2020). Note that the extracted travel time is shorter than the nominal SEIS sam-
 461 pling interval of 10 milliseconds, which illustrates the importance of the proposed recon-
 462 struction algorithm for the seismic analysis of the mole hammering data.

463 6 Conclusion

464 The high-frequency information of the HP³ hammering signal (frequencies above
 465 the nominal Nyquist frequency of 50 Hz) can be accurately recovered by the proposed
 466 reconstruction algorithm. Since the hammering time of the mole is uncorrelated with the
 467 sampling time of the seismometer, multiple realizations of approximately the same sig-
 468 nal are recorded, where each realization appears to be filtered with a Fourier-shift fil-
 469 ter. This allows for the recovery of the full-bandwidth signal by the application of the
 470 generalized sampling theorem. Since the signal is smoothly varying with depth as the
 471 mole slowly penetrates into the subsurface, we additionally make use of the Radon trans-
 472 form, which allows us to account for the resulting slope in the 2D signal. The maximum
 473 rate of change of the signal with depth is prescribed by the lowest propagation veloci-
 474 ties in the Martian ground, defining a limited area in the Radon transform domain where
 475 the signal has support. Reconstruction is then achieved by finding the sparsest set of Radon
 476 coefficients in this area that fit the data within the noise, allowing us to unwrap several
 477 orders of aliasing. We have demonstrated that this approach is robust also in the pres-
 478 ence of high levels of random noise due to the inherent properties of the Radon trans-
 479 form.

480 The proposed reconstruction algorithm could be adapted to similar problems of re-
 481 peated and only smoothly varying aliased and (quasi-)randomly sampled signals in sit-
 482 uations where sufficiently dense sampling along one dimension is not possible.

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Figure 1.

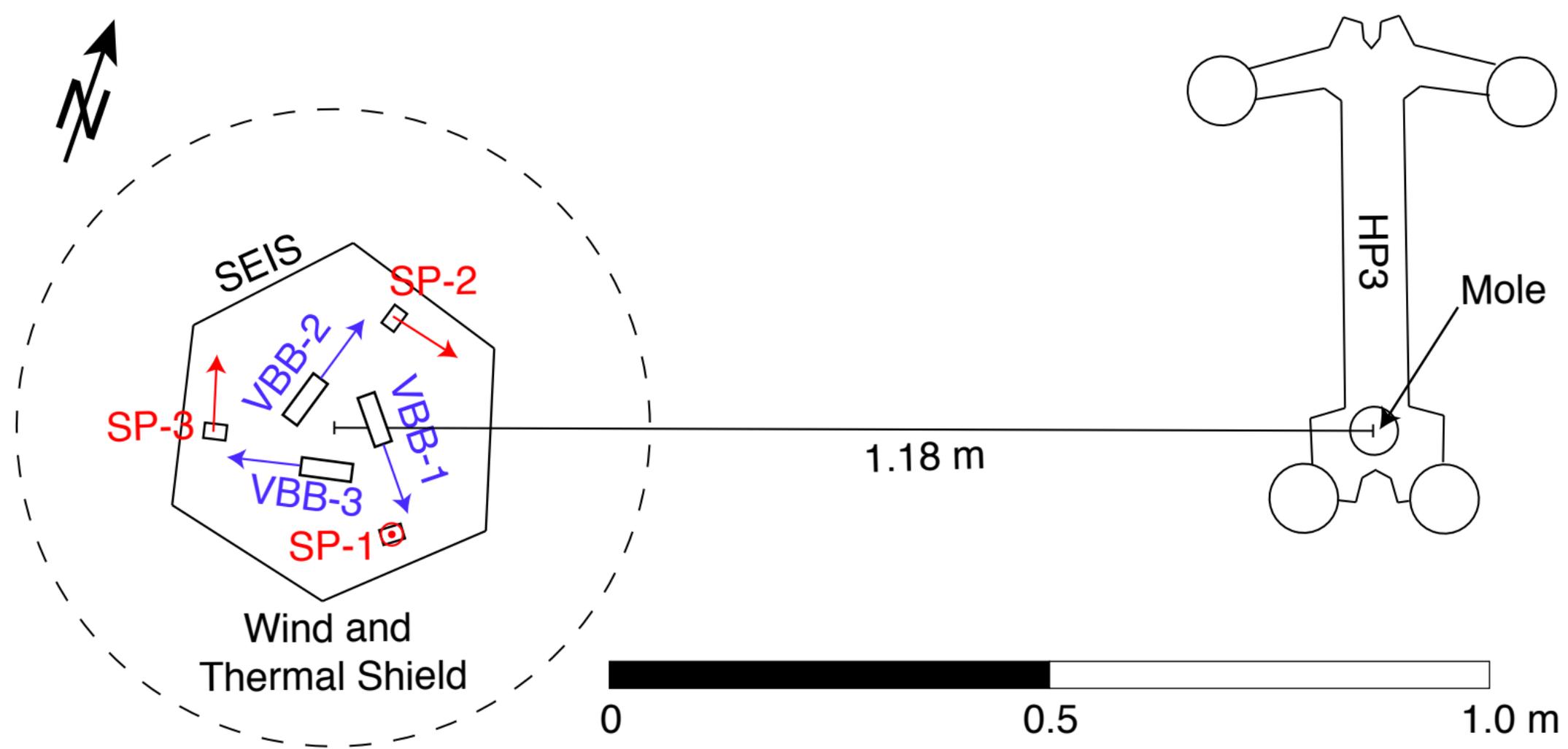
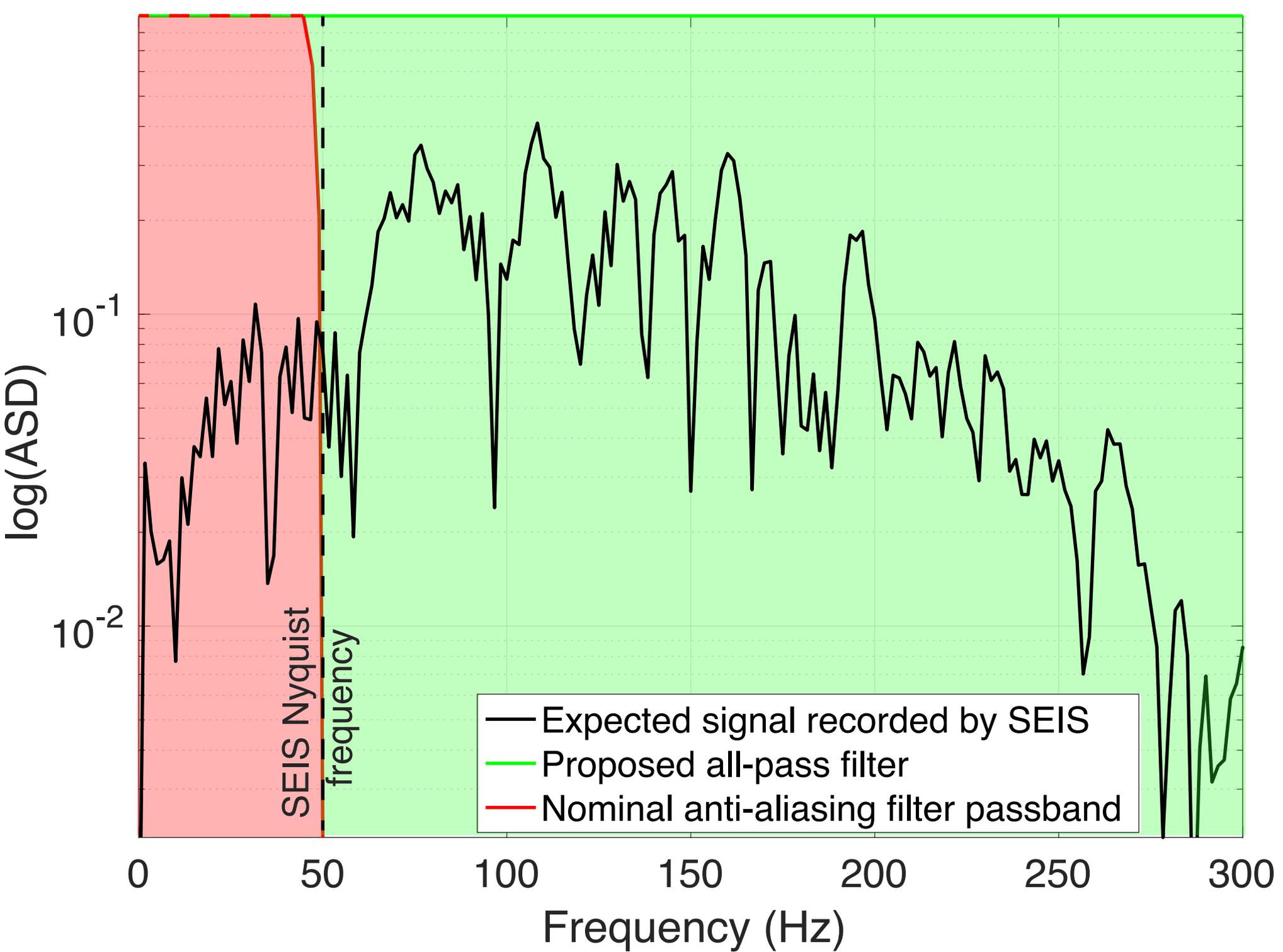


Figure 2.



- Expected signal recorded by SEIS
- Proposed all-pass filter
- Nominal anti-aliasing filter passband

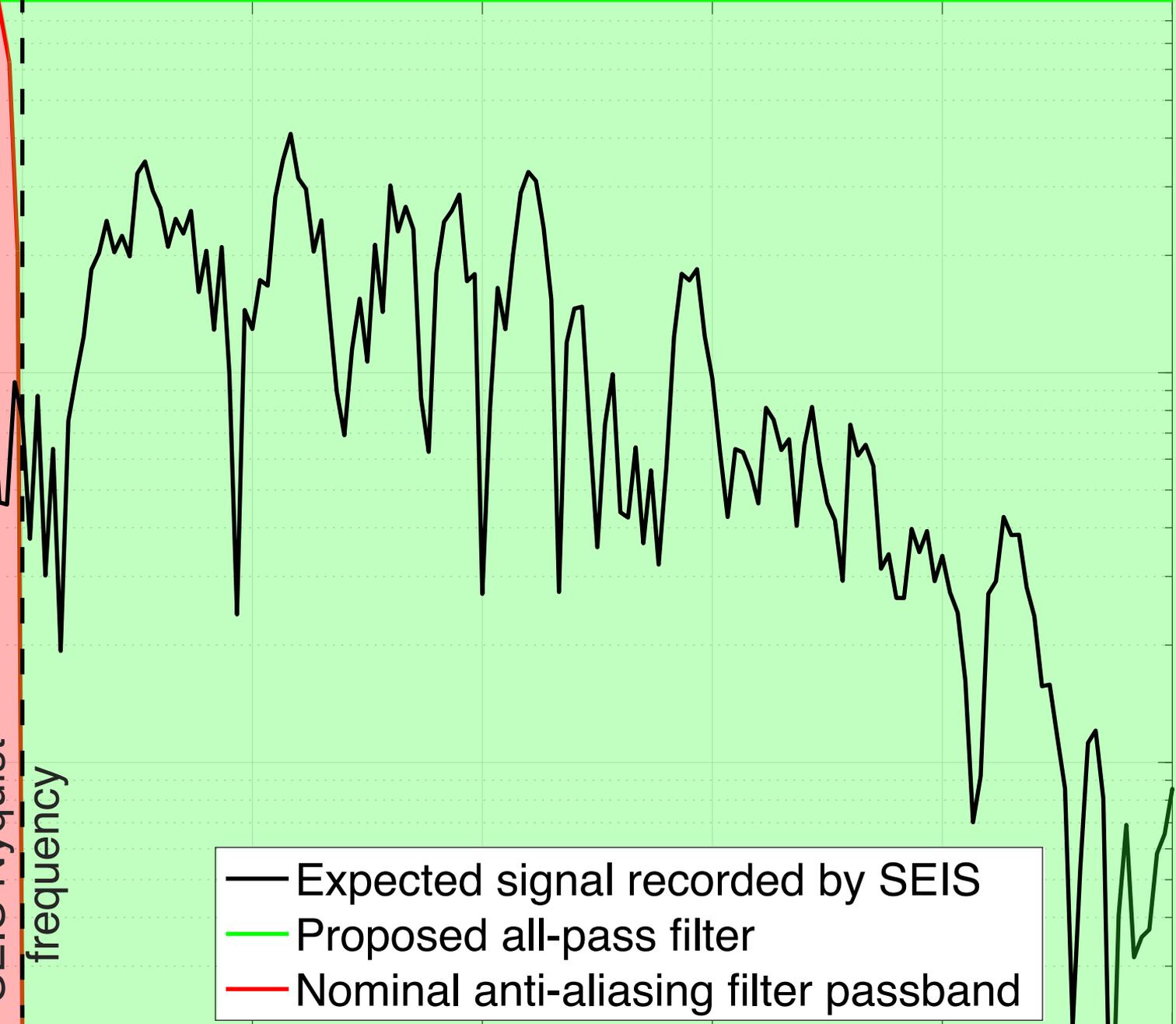
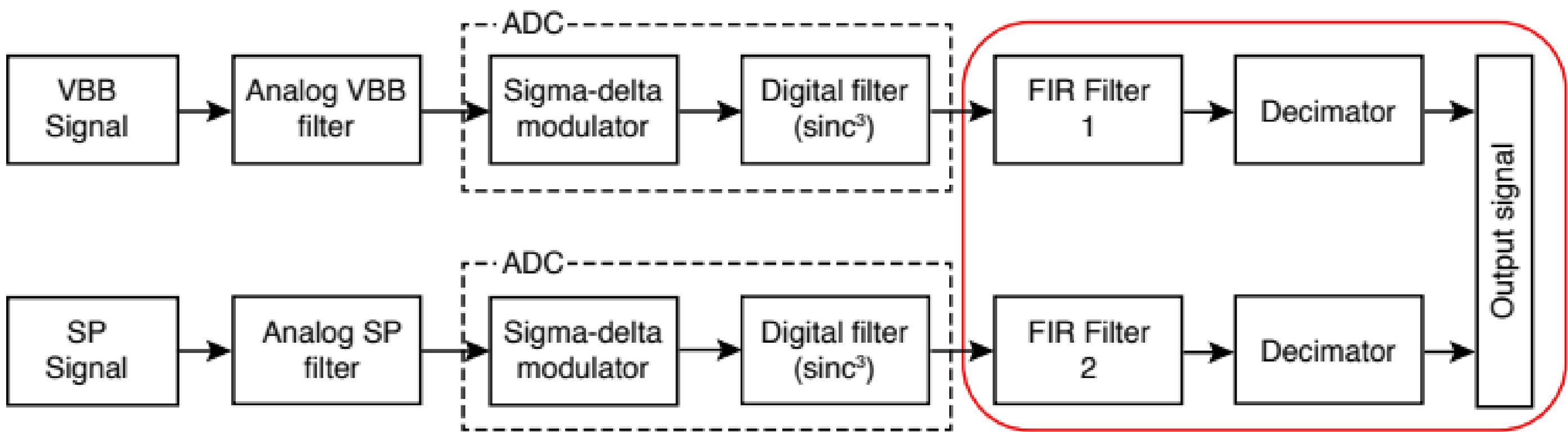


Figure 3.



Analog

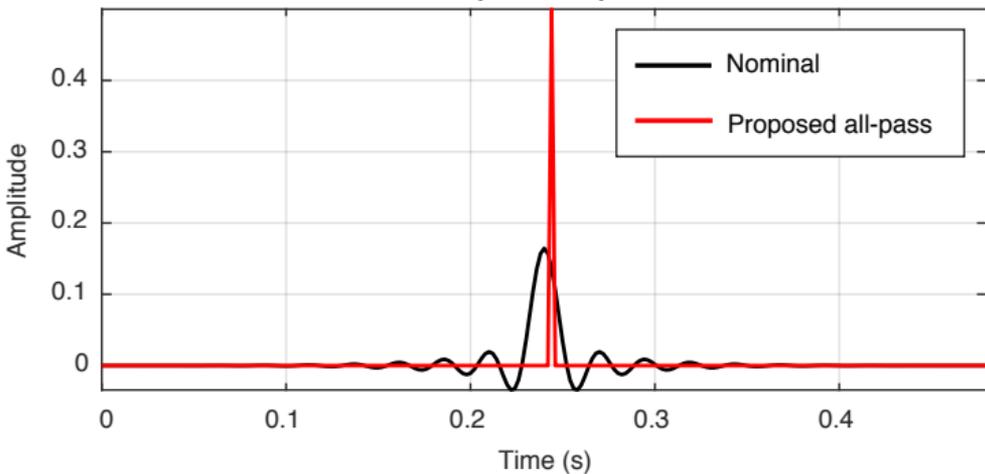
32'000 sps

500 sps

100 sps

Figure 4.

Impulse Response



Frequency Response

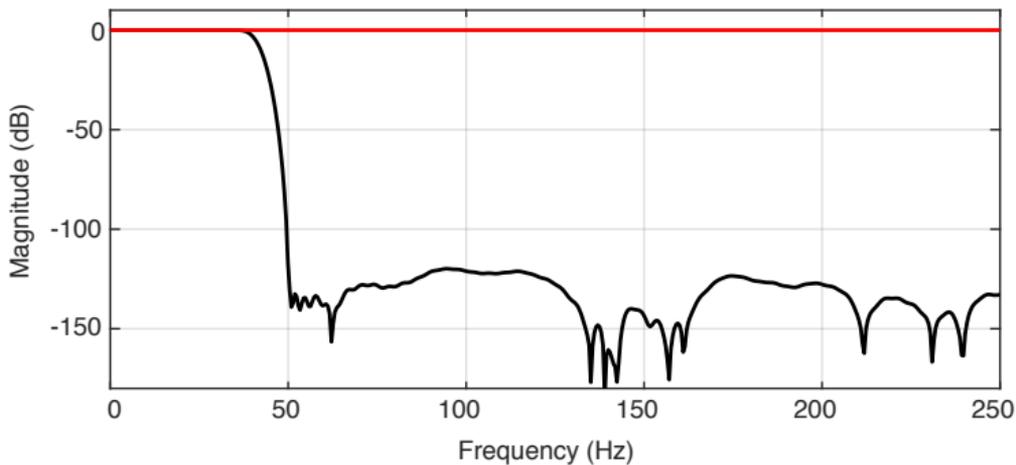


Figure 5.

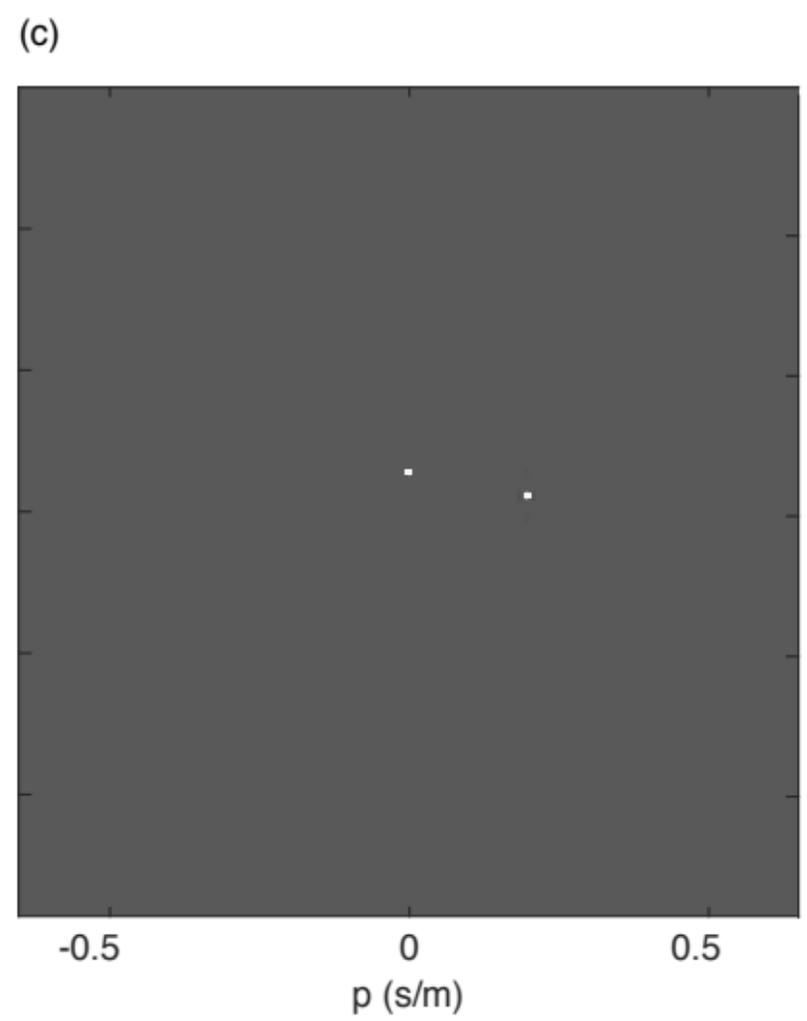
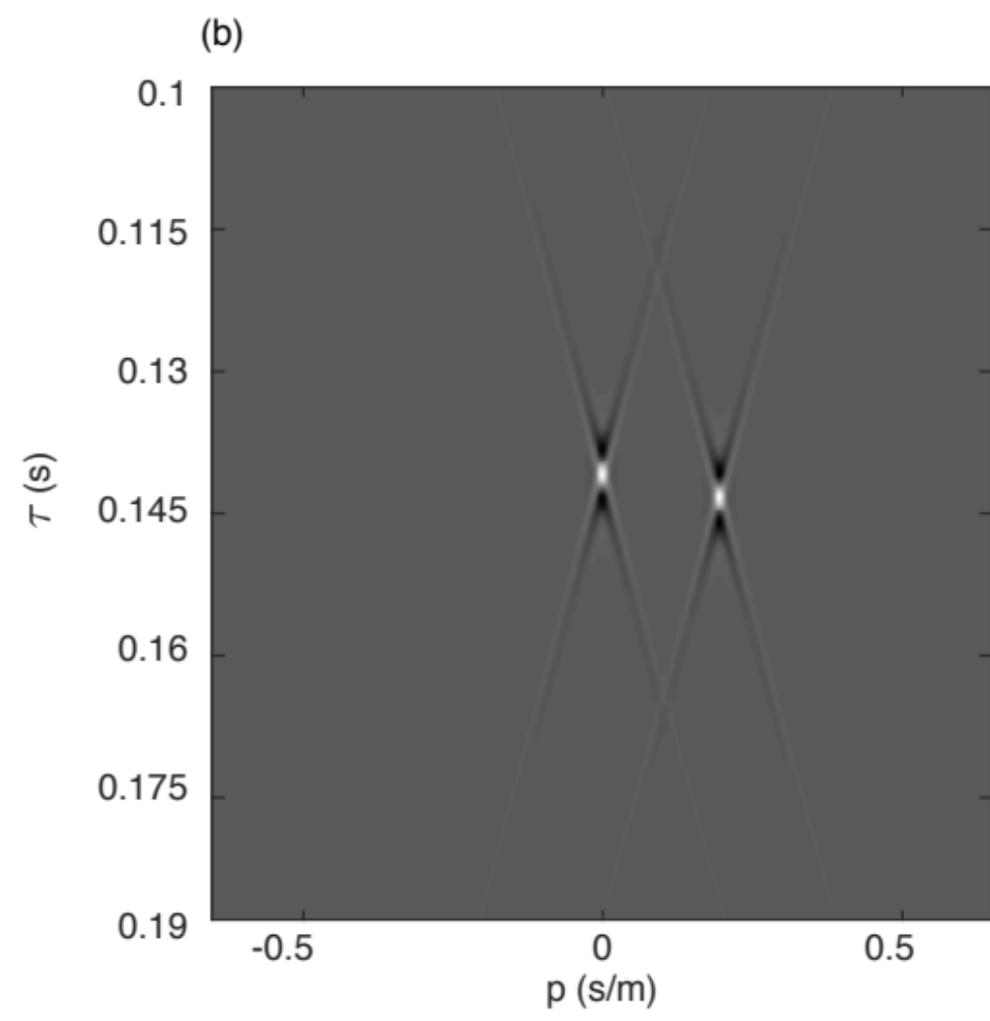
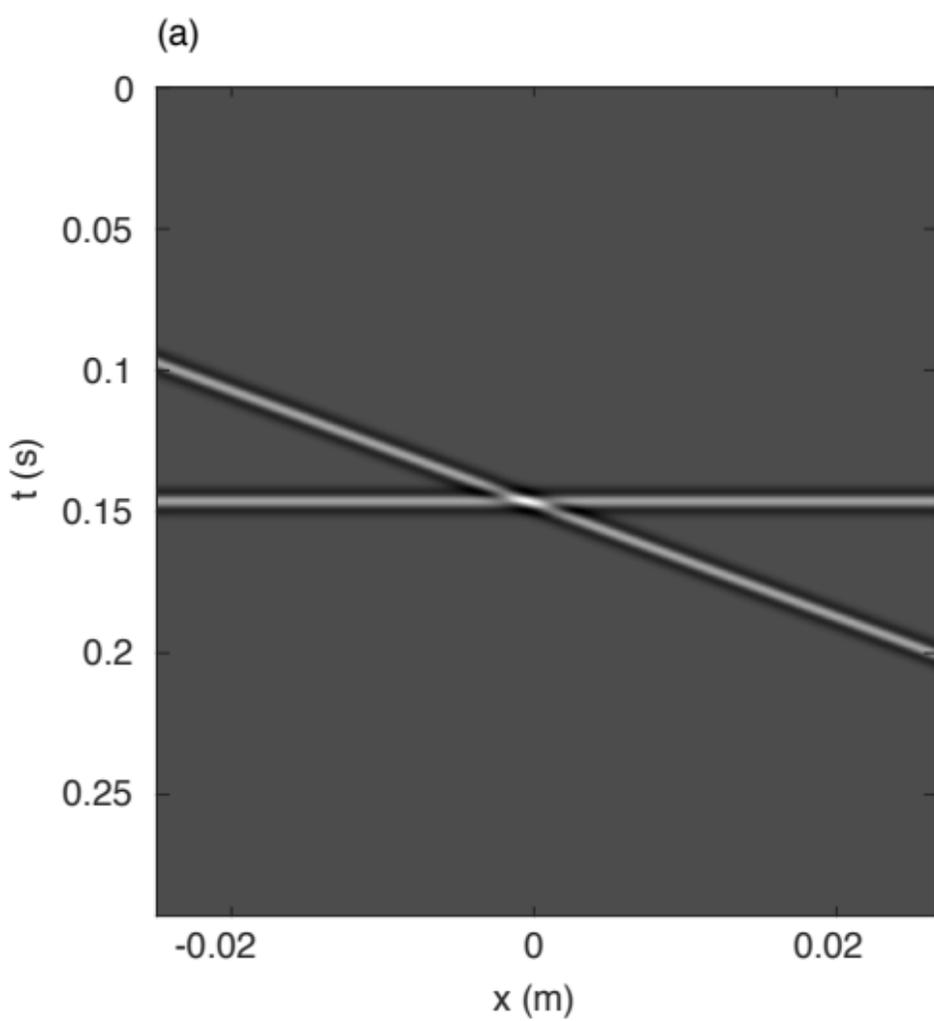


Figure 6.

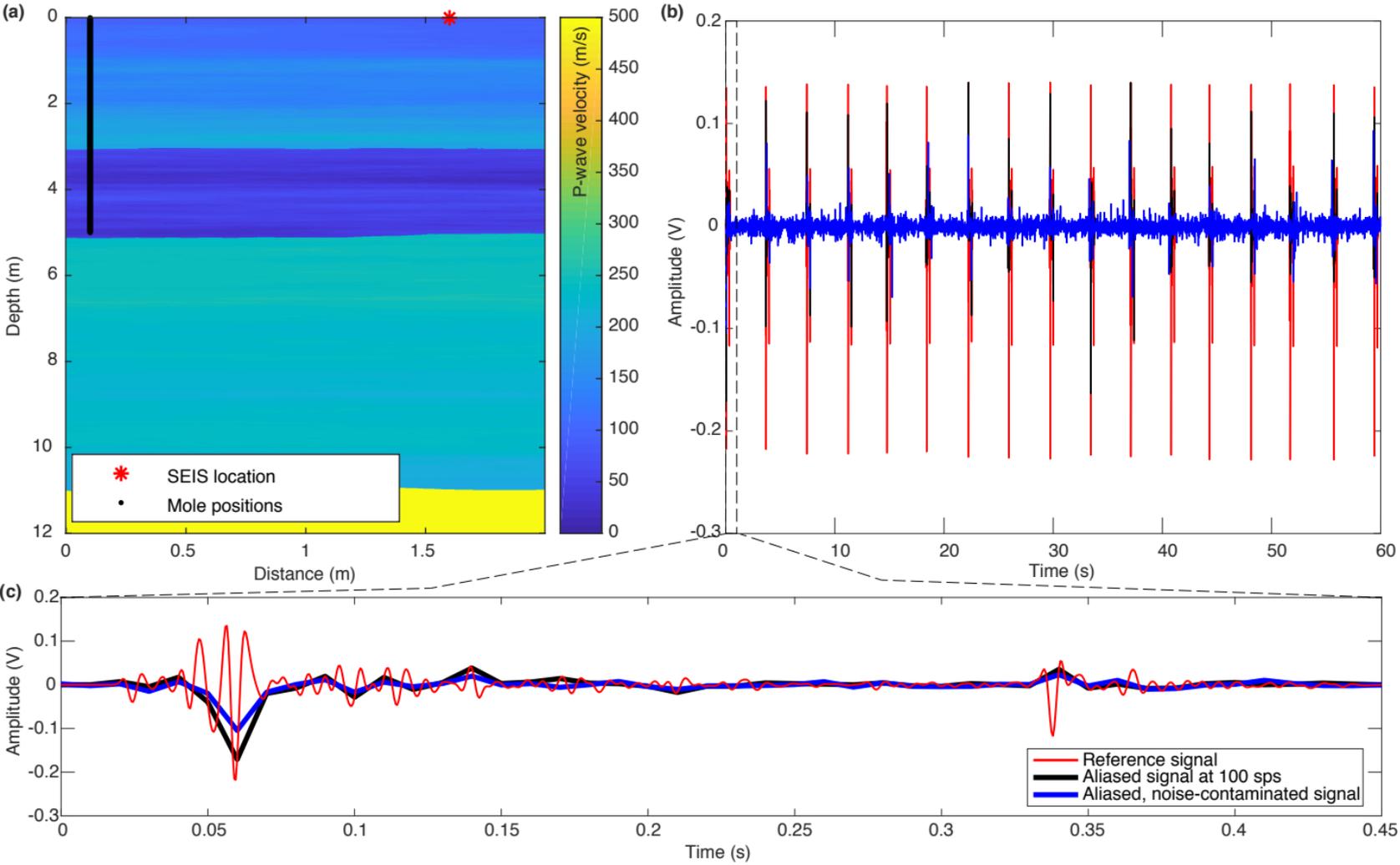


Figure 7.

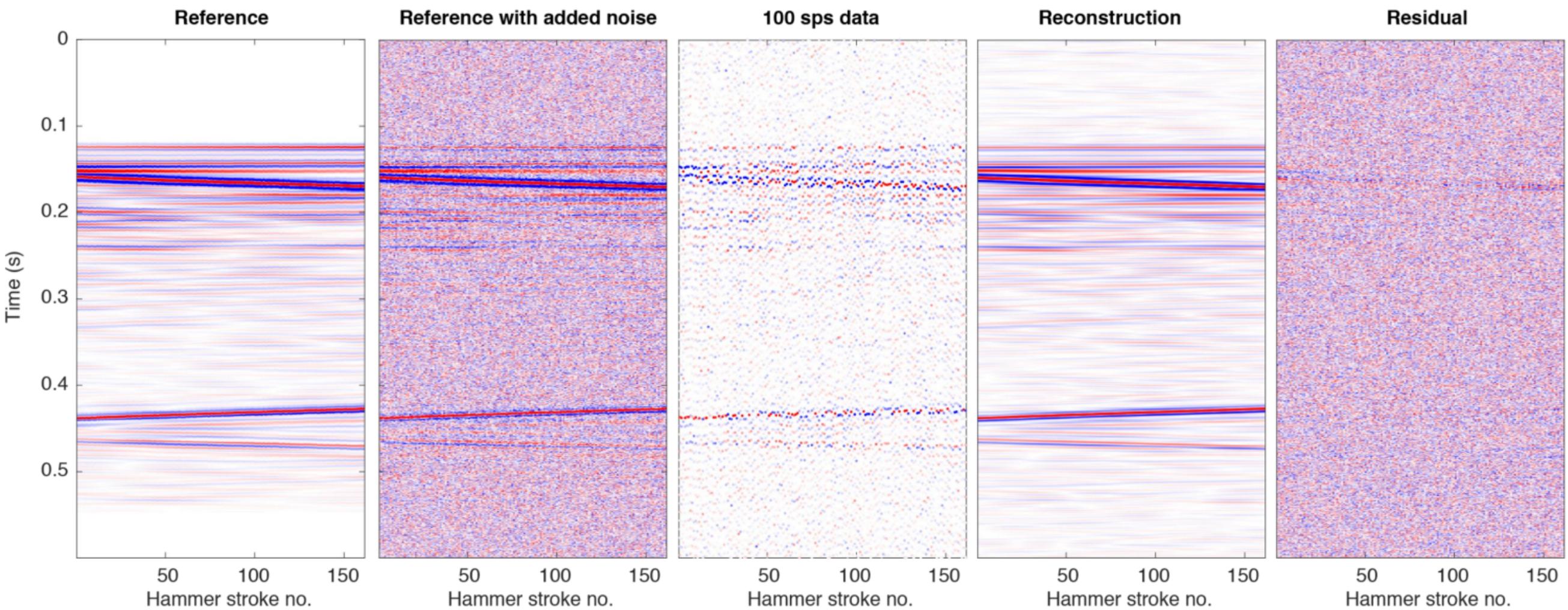


Figure 8.

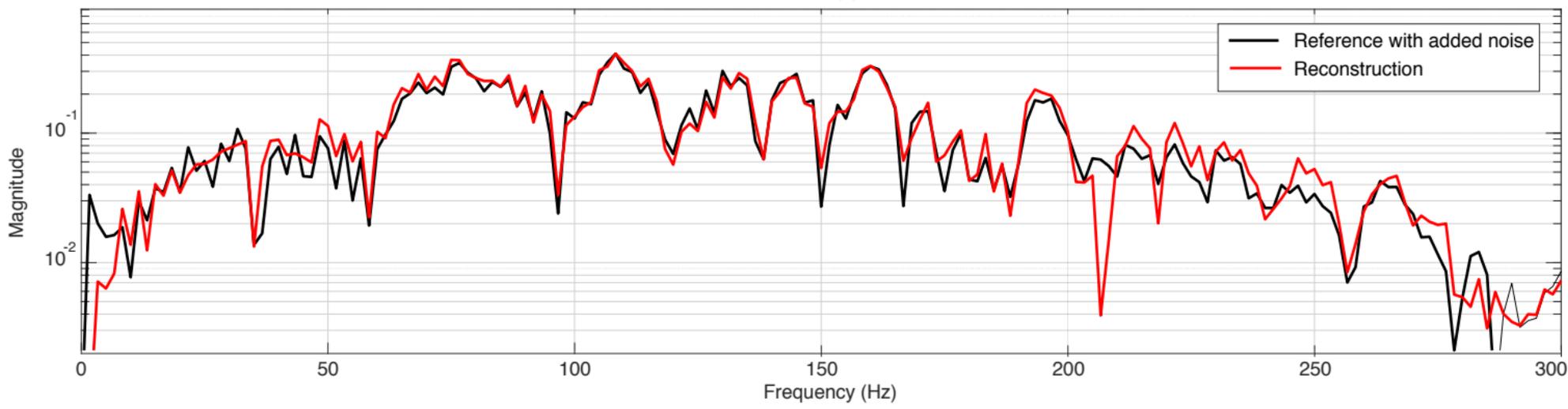
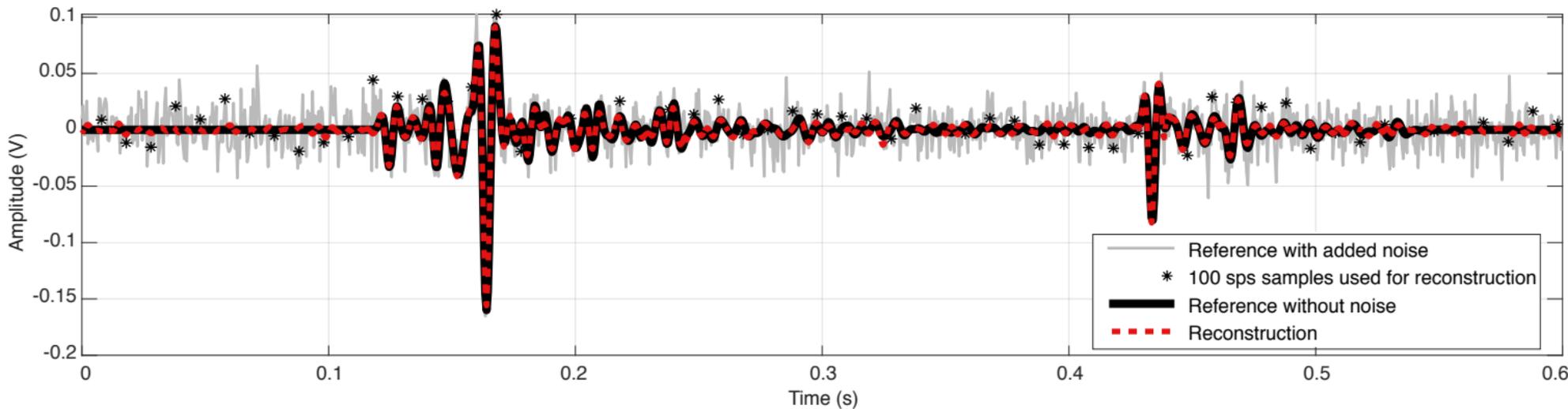


Figure 9.

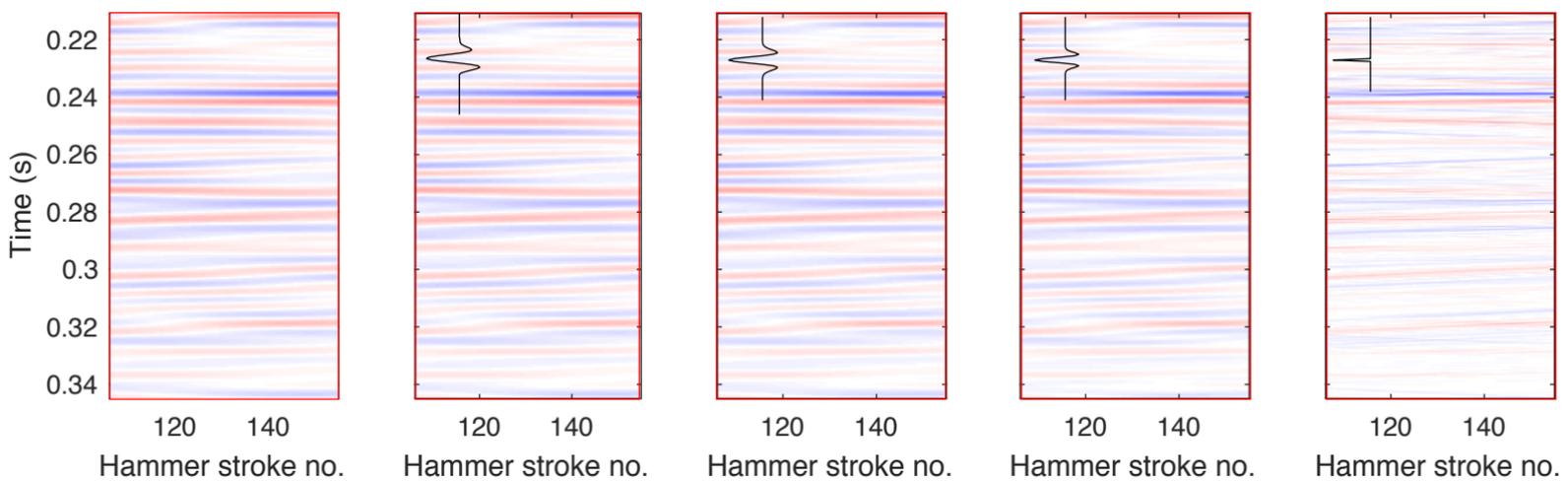
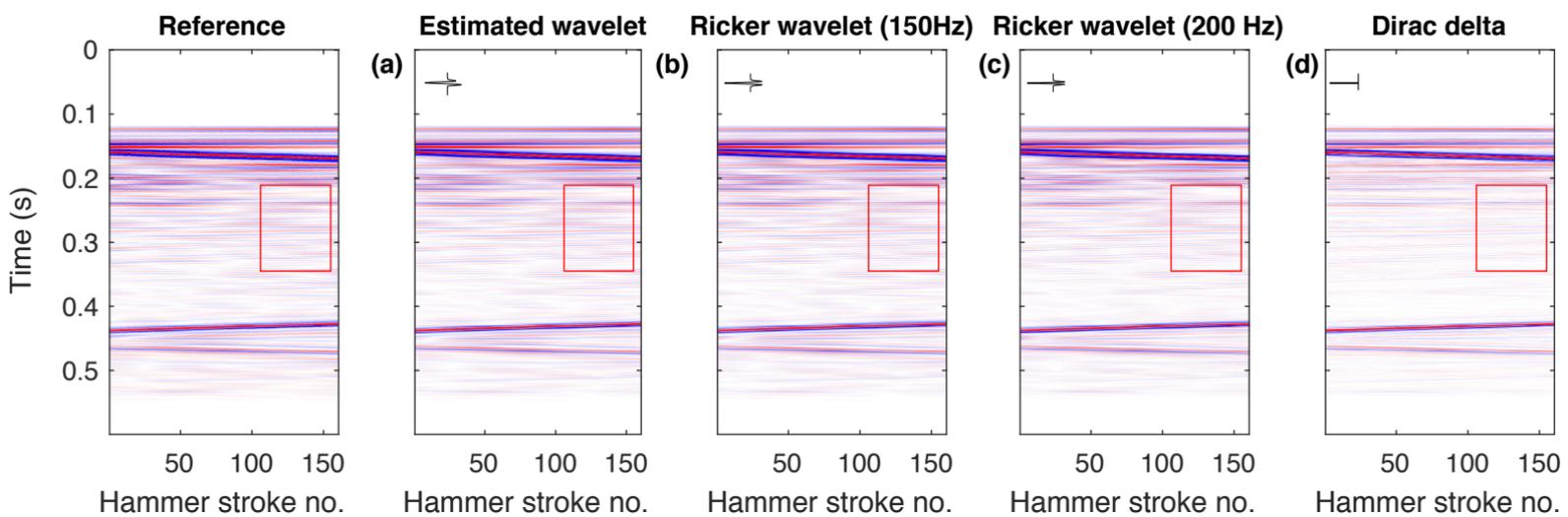


Figure 10.

